

Research Article

Robust H_∞ Filtering for Uncertain Discrete-Time Fuzzy Stochastic Systems with Sensor Nonlinearities and Time-Varying Delay

**Mingang Hua,^{1,2,3} Pei Cheng,⁴ Juntao Fei,¹
Jiayong Zhang,⁵ and Junfeng Chen¹**

¹ College of Computer and Information, Hohai University, Changzhou 213022, China

² Changzhou Key Laboratory of Sensor Networks and Environmental Sensing, Changzhou 213022, China

³ Jiangsu Key Laboratory of Power Transmission and Distribution Equipment Technology, Changzhou 213022, China

⁴ School of Mathematical Sciences, Anhui University, Hefei 230601, China

⁵ Department of Mathematics and Physics, Hohai University, Changzhou 213022, China

Correspondence should be addressed to Mingang Hua, mghua@yahoo.cn

Received 10 October 2012; Revised 7 December 2012; Accepted 11 December 2012

Academic Editor: Hak-Keung Lam

Copyright © 2012 Mingang Hua et al. This is an open access article distributed under the Creative Commons Attribution License, which permits unrestricted use, distribution, and reproduction in any medium, provided the original work is properly cited.

The robust filtering problem for a class of uncertain discrete-time fuzzy stochastic systems with sensor nonlinearities and time-varying delay is investigated. The parameter uncertainties are assumed to be time varying norm bounded in both the state and measurement equations. By using the Lyapunov stability theory and some new relaxed techniques, sufficient conditions are proposed to guarantee the robustly stochastic stability with a prescribed H_∞ performance level of the filtering error system for all admissible uncertainties, sensor nonlinearities, and time-varying delays. These conditions are dependent on the lower and upper bounds of the time-varying delays and are obtained in terms of a linear matrix inequality (LMI). Finally, two simulation examples are provided to illustrate the effectiveness of the proposed methods.

1. Introduction

Fuzzy systems in the Takagi-Sugeno (T-S) model can represent a lot of complex nonlinear systems in [1–3]. By using a T-S fuzzy plant model, one can describe a nonlinear system as a weighted sum of some simple linear subsystems. This fuzzy model is described by a family of fuzzy if-then rules that represent local linear input/output relations of the system. The overall fuzzy model of the system is achieved by smoothly blending these local linear models together through membership functions. Consequently, stability, control, and filtering problems for T-S fuzzy systems have attracted considerable attention, and many important results have been reported in [4–9].

It has been known that time delays are one of the main causes of instability and poor performance of control systems in [10]. Analysis and synthesis of such systems are of both theoretical and practical importance. Therefore, the study of time-delay systems has attracted great attention over the past few years in [11, 12]. For continuous-time systems, the obtained results can be generally classified into two types: delay-independent and delay-dependent ones. It has been understood that the latter is generally less conservative since the size of delays is considered, especially when time delays are small. Compared with continuous-time systems with time-varying delays, the discrete-time counterpart receives relatively less attention. See, for example, [13–16] and references therein.

In the past few years, much research effort has been paid to the control and filtering problems for nonlinear systems that have been widely applied in many fields such as communication network, image processing, and mobile robot localization [17, 18]. And the filtering for nonlinear stochastic systems has been of great interest since stochastic modeling has come to play an important role in many branches of engineering applications [19–22], particularly for discrete-time stochastic systems, so far, a number of important results have been reported for linear discrete-time stochastic systems. The exponential filtering problem is studied for discrete time-delay stochastic systems with Markovian jump parameters and missing measurements in [23]. The robust fault detection filter problem for fuzzy $It\hat{o}$ stochastic systems is studied in [24]. The problem of robust H_∞ filtering for uncertain discrete-time stochastic systems with time-varying delays is considered in [25]. For nonlinear discrete-time stochastic systems, the filtering problem for a class of nonlinear discrete-time stochastic systems with state delays is considered in [26]. The robust H_∞ filtering problem for a class of nonlinear discrete time-delay stochastic systems is considered in [27]. The H_∞ filtering problem for a general class of nonlinear discrete-time stochastic systems with randomly varying sensor delays is considered in [28]. Recently, the filtering problem for discrete-time fuzzy stochastic systems with sensor nonlinearities is considered in [29]. And the problem of H_∞ filtering for discrete-time Takagi-Sugeno (T-S) fuzzy $It\hat{o}$ stochastic systems with time-varying delay is studied in [30]. In [27, 29, 31, 32], the nonlinearity for filtering problem of systems was assumed to satisfy sensor nonlinearities, which may included actuator saturation and sensor saturation. It is worth mentioning that, although the system in [29] is with sensor nonlinearities, the proposed filter design approach do not consider the discrete-time fuzzy stochastic systems with time delay, which is not applicable to stochastic delay systems. And in [30], the stochastic systems do not contain the sensor nonlinearities. To the best of our knowledge, no results on H_∞ filtering for the uncertain discrete-time fuzzy stochastic systems with both sensor nonlinearities and time-varying delay are available in the literature, which motivates the present study.

Motivated by the works in [25, 27, 29, 30], in this paper, a delay-dependent H_∞ performance analysis result is established for filtering error systems. A new uncertain discrete-time fuzzy stochastic systems model is proposed, and a new different Lyapunov functional is then employed to deal with systems with sensor nonlinearities and time-varying delay. As a result, the H_∞ filters are designed in terms of linear matrix inequalities (LMIs). The resulting filters can ensure that the filtering error system is robustly stochastic stable and the estimation error is bounded by a prescribed level for all possible bounded energy disturbances. Finally, two examples are given to show the effectiveness of the proposed method.

Throughout this paper, \mathbb{R}^n denotes the n -dimensional Euclidean space, and $\mathbb{R}^{n \times m}$ is the set of $n \times m$ real matrices. I is the identity matrix. $|\cdot|$ denotes Euclidean norm for vectors and $\|\cdot\|$ denotes the spectral norm of matrices. N denotes the set of all natural numbers, that is, $N = \{0, 1, 2, \dots\}$. $(\Omega, \mathcal{F}, \{\mathcal{F}_k\}_{k \in N}, \mathcal{D})$ is a complete probability space with a filtration $\{\mathcal{F}_k\}_{k \in N}$ satisfying the usual conditions. M^T stands for the transpose of the matrix M . For

symmetric matrices X and Y , the notation $X > Y$ (resp., $X \geq Y$) means that the $X - Y$ is positive definite (resp., positive semidefinite). * denotes a block that is readily inferred by symmetry. $E\{\cdot\}$ stands for the mathematical expectation operator with respect to the given probability measure \mathcal{P} .

2. Problem Description

Consider a class of uncertain discrete-time stochastic systems that can be approximated by the following time-delay T-S fuzzy stochastic model with r plant rules.

Plant rule i :
if $\theta_1(k)$ is η_1^i and ... and $\theta_p(k)$ is η_p^i ,
then

$$\begin{aligned} x(k+1) &= (A_i + \Delta A_i(k))x(k) + (A_{di} + \Delta A_{di}(k))x(k - \tau(k)) + (B_i + \Delta B_i(k))v(k) \\ &\quad + [(E_i + \Delta E_i(k))x(k) + (E_{di} + \Delta E_{di}(k))x(k - \tau(k)) + (G_i + \Delta G_i(k))v(k)]w(k), \\ y(k) &= \phi(C_i(k))x(k) + (C_{di} + \Delta C_{di}(k))x(k - \tau(k)) + (D_i + \Delta D_i(k))v(k), \\ z(k) &= L_i x(k), \quad i = 1, 2, \dots, r, \end{aligned} \quad (2.1)$$

where η_j^i is the fuzzy set. $\theta(t) = [\theta_1(t), \theta_2(t), \dots, \theta_p(t)]^T$ is the premise variable vector, $x(k) \in \mathbb{R}^n$ is the state vector, $y(k) \in \mathbb{R}^g$ is the measurable output vector, $z(k) \in \mathbb{R}^r$ is the state combination to be estimated, $w(k)$ is a real scalar process on a probability space $(\Omega, \mathcal{F}, \mathcal{P})$ relative to an increasing family $(\mathcal{F}_k)_{k \in \mathbb{N}}$ of σ -algebra $\mathcal{F}_k \subset \mathcal{F}$ generated by $(w(k))_{k \in \mathbb{N}}$, and \mathbb{N} is the set of natural numbers. The stochastic process $\{w(k)\}$ is independent, which is assumed to satisfy

$$E\{w(k)\} = 0, \quad E\{w^2(k)\} = 1, \quad E\{w(i)w(j)\} = 0 \quad (i \neq j), \quad (2.2)$$

where the stochastic variables $w(0), w(1), w(2), \dots$ are assumed to be mutually independent. The exogenous disturbance signal $v(k) \in \mathbb{R}^p$ is assumed to belong to $L_{e2}([0, \infty), \mathbb{R}^p)$, and is \mathcal{F}_{k-1} measurable for all $k \in \mathbb{N}$, where $L_{e2}([0, \infty), \mathbb{R}^p)$ denotes the space of k -dimensional nonanticipatory square summable stochastic process $f(\cdot) = (f(k))_{k \in \mathbb{N}}$ on \mathbb{N} with respect to $\mathcal{F}_{k \in \mathbb{N}}$ satisfying

$$\|f\|_{e2}^2 = E\left\{\sum_{k=0}^{\infty} |f(k)|^2\right\} = \sum_{k=0}^{\infty} E\{|f(k)|^2\} < \infty. \quad (2.3)$$

The time-varying delay $\tau(k)$ satisfies

$$\tau_1 \leq \tau(k) \leq \tau_2, \quad (2.4)$$

where τ_1 and τ_2 are known positive integers representing the minimum and maximum delays, respectively.

In addition, $A_i, A_{di}, B_i, E_i, E_{di}, G_i, C_i, C_{di}, D_i$, and L_i are known real constant matrices, $\Delta A_i(k), \Delta A_{di}(k), \Delta B_i(k), \Delta E_i(k), \Delta E_{di}(k), \Delta G_i(k), \Delta C_{di}(k)$, and $\Delta D_i(k)$ are unknown matrices representing time-varying parameter uncertainties, and the admissible uncertainties are assumed to be modeled in the form

$$[\Delta A_i(k) \ \Delta E_i(k)] = [M_{1i} \ M_{2i}]F_{1i}(k)N_{1i}, \quad (2.5)$$

$$[\Delta A_{di}(k) \ \Delta E_{di}(k) \ \Delta C_{di}(k)] = [M_{3i} \ M_{4i} \ M_{5i}]F_{2i}(k)N_{2i}, \quad (2.6)$$

$$[\Delta B_i(k) \ \Delta G_i(k) \ \Delta D_i(k)] = [M_{6i} \ M_{7i} \ M_{8i}]F_{3i}(k)N_{3i}, \quad (2.7)$$

where $M_{1i}, M_{2i}, M_{3i}, M_{4i}, M_{5i}, M_{6i}, M_{7i}$, and M_{8i} are known constant matrices and $F_{1i}(k), F_{2i}(k)$, and $F_{3i}(k)$ are unknown time-varying matrices with Lebesgue measurable elements bounded by

$$F_{1i}^T(k)F_{1i}(k) \leq I, \quad F_{2i}^T(k)F_{2i}(k) \leq I, \quad F_{3i}^T(k)F_{3i}(k) \leq I, \quad \forall k. \quad (2.8)$$

The defuzzified output of the T-S fuzzy system (2.1) is inferred as follows:

$$\begin{aligned} x(k+1) &= \sum_{i=1}^r h_i(\theta(k)) \{ (A_i + \Delta A_i(k))x(k) + (A_{di} + \Delta A_{di}(k))x(k - \tau(k)) + (B_i + \Delta B_i(k))v(k) \\ &\quad + [(E_i + \Delta E_i(k))x(k) + (E_{di} + \Delta E_{di}(k))x(k - \tau(k)) \\ &\quad + (G_i + \Delta G_i(k))v(k)]w(k) \}, \\ y(k) &= \sum_{i=1}^r h_i(\theta(k)) \{ \phi(C_i(k))x(k) + (C_{di} + \Delta C_{di}(k))x(k - \tau(k)) + (D_i + \Delta D_i(k))v(k) \}, \\ z(k) &= \sum_{i=1}^r h_i(\theta(k))L_i x(k), \end{aligned} \quad (2.9)$$

where $h_i(\theta(k)) = \mu_i(\theta(k)) / \sum_{j=1}^r h_j(\theta(k))$, and $\mu_i(\theta(k)) = \prod_{j=1}^p \eta_j^i(\theta_j(k))$, in which $\eta_j^i(\theta_j(k))$ is the grade of membership of $\eta_j(k)$ in η_j^i . According to the theory of fuzzy sets, we have $\mu_i(\theta(k)) \geq 0, (i = 1, 2, \dots, r)$ and $\sum_{i=1}^r h_i(\theta(k)) > 0$. Therefore, it is implied that $h_i(\theta(k)) \geq 0, (i = 1, 2, \dots, r)$ and $\sum_{i=1}^r h_i(\theta(k)) = 1$.

For some given diagonal matrices $K_1 \geq 0$ and $K_2 \geq 0$ with $K_2 > K_1$, the nonlinear vector functions $\phi \in [K_1, K_2]$, is assumed to represent *sensor nonlinearities* and satisfies the following sensor condition [29]:

$$(\phi(\xi) - K_1\xi)^T (\phi(\xi) - K_2\xi) \leq 0, \quad \forall \xi \in \mathbb{R}^q. \quad (2.10)$$

Further, the nonlinear function $\phi(\xi)$ can be decomposed into a linear and a nonlinear part as follows:

$$\phi(\xi) = \phi_s(\xi) + K_1\xi, \quad (2.11)$$

where the nonlinearity $\phi_s(\xi)$ belongs to the set Φ_s given by

$$\Phi_s = \left\{ \phi_s : \phi_s^T(\xi)(\phi_s(\xi) - K\xi) \leq 0 \right\} \quad (2.12)$$

with $K = K_2 - K_1 > 0$.

We consider the following fuzzy filter for the estimation of z_k :

$$\begin{aligned} \hat{x}(k+1) &= \sum_{i=1}^r h_i(\theta(k)) \{ A_{fi} \hat{x}(k) + B_{fi} y(k) \}, \\ \hat{z}(k) &= \sum_{i=1}^r h_i(\theta(k)) L_i \hat{x}(k), \end{aligned} \quad (2.13)$$

where $\hat{x}(k) \in \mathbb{R}^n$, $\hat{z}(k) \in \mathbb{R}^r$, and the matrices A_{fi} and B_{fi} ($i = 1, 2, \dots, r$) are to be determined.

Remark 2.1. Similar to [27, 29], the sensor nonlinearities satisfying (2.10) are also considered in this paper. Here, there exists the nonlinear function $\phi(\xi)$ in the system (2.1), which is called *sensor nonlinearities*. It is noted that in the previous filter, the matrix L_i is assumed to be constant in order to avoid more verbosely mathematical derivation.

Defining $\tilde{x}(k) = x(k) - \hat{x}(k)$, $\tilde{z}(k) = z(k) - \hat{z}(k)$, $e(k) = [x^T(k) \ \tilde{x}^T(k)]^T$, and augmenting the model (2.9) to include the states of the filter (2.13), we obtain the following filtering error system:

$$\begin{aligned} e(k+1) &= \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k)) h_j(\theta(k)) \left\{ \left(\bar{A}_{ij} + \Delta \bar{A}_{ij}(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) \right. \\ &\quad + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) H e(k - \tau(k)) + \left(\bar{B}_{ij} + \Delta \bar{B}_{ij}(k) \right) v(k) \\ &\quad + \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) H e(k - \tau(k)) \right. \\ &\quad \left. \left. + \left(\bar{G}_i + \Delta \bar{G}_i(k) \right) v(k) \right] w(k) \right\}, \end{aligned} \quad (2.14)$$

$$\tilde{z}(k) = \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k)) h_j(\theta(k)) \bar{L}_{ij} e(k), \quad (2.15)$$

where

$$\begin{aligned}
\bar{A}_{ij} &= \begin{bmatrix} A_i & 0 \\ A_i - A_{fj} - B_{fj}K_1C_i & A_{fj} \end{bmatrix}, & \bar{B}_{fj} &= \begin{bmatrix} 0 \\ -B_{fj} \end{bmatrix}, & \bar{C}_i &= [C_i \ 0], \\
\bar{A}_{dij} &= \begin{bmatrix} A_{di} \\ A_{di} - B_{fj}C_{di} \end{bmatrix}, & \bar{B}_{ij} &= \begin{bmatrix} B_i \\ B_i - B_{fj}D_i \end{bmatrix}, & \bar{E}_i &= \begin{bmatrix} E_i & 0 \\ E_i & 0 \end{bmatrix}, \\
\bar{E}_{di} &= \begin{bmatrix} E_{di} \\ E_{di} \end{bmatrix}, & \bar{G}_i &= \begin{bmatrix} G_i \\ G_i \end{bmatrix}, & \bar{L}_{ij} &= [0 \ L_i - L_j], & H &= [I \ 0], \\
\Delta \bar{A}_i(k) &= \bar{M}_{A_i} F_{1i}(k) \bar{N}_{1i}, & \Delta \bar{A}_{dij}(k) &= \bar{M}_{A_{dij}} F_{2i}(k) N_{2i}, & \Delta \bar{B}_{ij}(k) &= \bar{M}_{B_{ij}} F_{3i}(k) N_{3i}, \\
\Delta \bar{E}_i(k) &= \bar{M}_{E_i} F_{1i}(k) \bar{N}_{1i}, & \Delta \bar{E}_{di}(k) &= \bar{M}_{E_{di}} F_{2i}(k) N_{2i}, & \Delta \bar{G}_i(k) &= \bar{M}_{G_i} F_{3i}(k) N_{3i}, \\
\bar{M}_{A_i} &= \begin{bmatrix} M_{1i} \\ M_{1i} \end{bmatrix}, & \bar{M}_{A_{dij}} &= \begin{bmatrix} M_{3i} \\ M_{3i} - B_{fj}M_{5i} \end{bmatrix}, & \bar{M}_{B_{ij}} &= \begin{bmatrix} M_{6i} \\ M_{6i} - B_{fj}M_{8i} \end{bmatrix}, \\
\bar{M}_{E_i} &= \begin{bmatrix} M_{2i} \\ M_{2i} \end{bmatrix}, & \bar{M}_{E_{di}} &= \begin{bmatrix} M_{4i} \\ M_{4i} \end{bmatrix}, & \bar{M}_{G_i} &= \begin{bmatrix} M_{7i} \\ M_{7i} \end{bmatrix}, & \bar{N}_{1i} &= [N_{1i} \ 0].
\end{aligned} \tag{2.16}$$

The H_∞ filtering problem to be addressed in this paper can be formulated as follows. Given discrete-time fuzzy stochastic systems (2.9), a prescribed level of noise attenuation $\gamma > 0$, and any $\phi \in [K_1 \ K_2]$, find a suitable filter in the form of (2.13) such that the following requirements are satisfied.

- (1) The filtering error system (2.14)-(2.15) with $v(k) = 0$ is said to be robustly stochastic stable if there exists a scalar $c > 0$ such that for all admissible uncertainties satisfying (2.5)–(2.8)

$$\mathbb{E} \left\{ \sum_{k=0}^{\infty} |x(k)|^2 \right\} \leq c \mathbb{E} \{ |x(0)|^2 \}, \tag{2.17}$$

where $x(k)$ denotes the solution of stochastic systems with initial state $x(0)$.

- (2) For the given disturbance attenuation level $\gamma > 0$ and under zero initial conditions for all $v(k) \in L_{e2}([0 \ \infty), \mathbb{R}^p)$, the performance index γ satisfies the following inequality:

$$\|z(k) - \hat{z}(k)\|_{e2} < \gamma \|v(k)\|_{e2}. \tag{2.18}$$

Before concluding this section, we introduce the following lemmas, which will be used in the derivation of our main results in the next section.

Lemma 2.2 (Gu et al. [33]). *Given constant matrices Γ_1 , Γ_2 , and Γ_3 with appropriate dimensions, where $\Gamma_1^T = \Gamma_1$ and $\Gamma_2^T = \Gamma_2 > 0$, then*

$$\Gamma_1 + \Gamma_3^T \Gamma_2^{-1} \Gamma_3 < 0 \tag{2.19}$$

if and only if

$$\begin{bmatrix} \Gamma_1 & \Gamma_3^T \\ \Gamma_3 & -\Gamma_2 \end{bmatrix} < 0, \quad \text{or} \quad \begin{bmatrix} -\Gamma_2 & \Gamma_3 \\ \Gamma_3^T & \Gamma_1 \end{bmatrix} < 0. \quad (2.20)$$

Lemma 2.3 (see [34]). For given matrices H , E , and $F(t)$ with $F^T(t)F(t) \leq I$ and scalar $\varepsilon > 0$, the following inequality holds:

$$HF(t)E + E^T F^T(t)H^T \leq \varepsilon HH^T + \varepsilon^{-1} E^T E. \quad (2.21)$$

3. Main Results

Theorem 3.1. Consider the uncertain discrete-time fuzzy stochastic systems in (2.1). A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system (2.14)-(2.15) is robustly stochastic stable with performance γ , if there exist real matrices $X > 0$, $Y > 0$, $Q > 0$, any matrices A_{Fj} , B_{Fj} , scalars $\varepsilon_{1i} > 0$, $\varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$ such that the following LMI is satisfied:

$$\begin{bmatrix} \Sigma_{11} & 0 & 0 & 0 & C_i^T K^T & A_i^T X & \Sigma_{17} & E_i^T X & E_i^T Y & 0 & 0 & 0 \\ * & \Sigma_{22} & 0 & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 & 0 \\ * & * & \Sigma_{33} & 0 & 0 & A_{di}^T X & \Sigma_{37} & E_{di}^T X & E_{di}^T Y & 0 & 0 & 0 \\ * & * & * & \Sigma_{44} & 0 & B_i^T X & \Sigma_{47} & G_i^T X & G_i^T Y & 0 & 0 & 0 \\ * & * & * & * & -2I & 0 & -B_{Fj}^T & 0 & 0 & 0 & 0 & 0 \\ * & * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{3i} & XM_{6i} \\ * & * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Sigma_{711} & \Sigma_{712} \\ * & * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{4i} & XM_{7i} \\ * & * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{4i} & YM_{7i} \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{1i} I & 0 & 0 \\ * & * & * & * & * & * & * & * & * & * & -\varepsilon_{2i} I & 0 \\ * & * & * & * & * & * & * & * & * & * & * & -\varepsilon_{3i} I \end{bmatrix} < 0, \quad (3.1)$$

where

$$\begin{aligned} \Sigma_{11} &= -X + (\tau_2 - \tau_1 + 1)Q + \varepsilon_{1i} N_{1i}^T N_{1i}, & \Sigma_{22} &= -Y + (L_i - L_j)^T (L_i - L_j), \\ \Sigma_{33} &= -Q + \varepsilon_{2i} N_{2i}^T N_{2i}, & \Sigma_{44} &= -\gamma^2 I + \varepsilon_{3i} N_{3i}^T N_{3i}, \\ \Sigma_{17} &= A_i^T Y - A_{Fj}^T - C_i^T K_1^T B_{Fj}^T, & \Sigma_{37} &= A_{di}^T Y - C_{di}^T B_{Fj}^T, & \Sigma_{47} &= B_i^T Y - D_i^T B_{Fj}^T, \\ \Sigma_{711} &= YM_{3i} - B_{Fj} M_{5i}, & \Sigma_{712} &= YM_{6i} - B_{Fj} M_{8i}. \end{aligned} \quad (3.2)$$

Moreover, if the previous condition is satisfied, an acceptable state-space realization of the H_∞ filter is given by

$$A_{fj} = Y^{-1} A_{Fj}, \quad B_{fj} = Y^{-1} B_{Fj}. \quad (3.3)$$

Proof. We first establish the condition of robustly stochastic stability for the filtering error system (2.14)-(2.15). It can be shown that LMI (3.1) implies

$$\begin{bmatrix} \Sigma_{11} & 0 & 0 & C_i^T K^T & A_i^T X & \Sigma_{17} & E_i^T X & E_i^T Y & 0 & 0 \\ * & -Y & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 \\ * & * & \Sigma_{33} & 0 & A_{di}^T X & \Sigma_{37} & E_{di}^T X & E_{di}^T Y & 0 & 0 \\ * & * & * & -2I & 0 & B_{Fj}^T & 0 & 0 & 0 & 0 \\ * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{3i} \\ * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Sigma_{711} \\ * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{4i} \\ * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{4i} \\ * & * & * & * & * & * & * & * & -\varepsilon_{1i}I & 0 \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{2i}I \end{bmatrix} < 0. \quad (3.4)$$

Define a matrix $\bar{P} > 0$ by

$$\bar{P} = \begin{bmatrix} X & 0 \\ 0 & Y \end{bmatrix}, \quad (3.5)$$

where $X > 0$ and $Y > 0$ satisfy the solvability of (3.1).

Then, it can be shown from (3.3) and (3.5) that LMI (3.4) can be rewritten as

$$\begin{bmatrix} \bar{\Sigma}_{11} & 0 & \bar{C}_i^T K^T & \bar{A}_{ij}^T \bar{P} & \bar{E}_i^T \bar{P} & 0 & 0 \\ * & \bar{\Sigma}_{22} & 0 & H^T \bar{A}_{dij}^T \bar{P} & H^T \bar{E}_{di}^T \bar{P} & 0 & 0 \\ * & * & -2I & \bar{B}_{fj}^T \bar{P} & 0 & 0 & 0 \\ * & * & * & -\bar{P} & 0 & \bar{P} \bar{M}_{Ai} & \bar{P} \bar{M}_{A_{dij}} \\ * & * & * & * & -\bar{P} & \bar{P} \bar{M}_{Ei} & \bar{P} \bar{M}_{E_{di}} \\ * & * & * & * & * & -\varepsilon_{1i}I & 0 \\ * & * & * & * & * & * & -\varepsilon_{2i}I \end{bmatrix} < 0, \quad (3.6)$$

where

$$\bar{\Sigma}_{11} = -\bar{P} + (\tau_2 - \tau_1 + 1)H^T QH + \varepsilon_{1i} \bar{N}_{1i}^T \bar{N}_{1i}, \quad \bar{\Sigma}_{22} = -H^T QH + \varepsilon_{2i} H^T N_{2i}^T N_{2i} H, \quad (3.7)$$

which is equivalent to the following inequality:

$$\begin{bmatrix} \tilde{\Sigma}_{11} & 0 & \bar{C}_i^T K^T & \bar{A}_{ij}^T \bar{P} & \bar{E}_i^T \bar{P} \\ * & -H^T QH & 0 & H^T \bar{A}_{dij}^T \bar{P} & H^T \bar{E}_{di}^T \bar{P} \\ * & * & -2I & \bar{B}_{fj}^T \bar{P} & 0 \\ * & * & * & -\bar{P} & 0 \\ * & * & * & * & -\bar{P} \end{bmatrix} + \varepsilon_{1i} \bar{N}_{1i}^T \bar{N}_{1i} + \varepsilon_{1i}^{-1} \bar{M}_{1i} \bar{M}_{1i}^T \quad (3.8)$$

$$+ \varepsilon_{2i} \bar{N}_{2i}^T \bar{N}_{2i} + \varepsilon_{2i}^{-1} \bar{M}_{2i} \bar{M}_{2i}^T < 0$$

with

$$\begin{aligned}\tilde{\Sigma}_{11} &= -\bar{P} + (\tau_2 - \tau_1 + 1)H^TQH, & \tilde{M}_{1i} &= \begin{bmatrix} 0 & 0 & 0 & \bar{M}_{A_{dij}}^T \bar{P} & \bar{M}_{E_{di}}^T \bar{P} \end{bmatrix}^T, \\ \tilde{N}_{1i} &= \begin{bmatrix} \bar{N}_{1i} & 0 & 0 & 0 & 0 \end{bmatrix}, & \tilde{M}_{2i} &= \begin{bmatrix} 0 & 0 & 0 & \bar{M}_{A_i}^T \bar{P} & \bar{M}_{E_i}^T \bar{P} \end{bmatrix}^T, \\ \tilde{N}_{2i} &= \begin{bmatrix} 0 & N_{2i}H & 0 & 0 & 0 \end{bmatrix}.\end{aligned}\quad (3.9)$$

On the other hand, it is noted that the following equality holds:

$$\begin{aligned}& \begin{bmatrix} \tilde{\Sigma}_{11} & 0 & \bar{C}_i^T K^T & (\bar{A}_{ij} + \Delta \bar{A}_i(k))^T \bar{P} & (\bar{E}_i + \Delta \bar{E}_i(k))^T \bar{P} \\ * & -H^TQH & 0 & H^T(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k))^T \bar{P} & H^T(\bar{E}_{di} + \Delta \bar{E}_{di}(k))^T \bar{P} \\ * & * & -2I & \bar{B}_{fj}^T \bar{P} & 0 \\ * & * & * & -\bar{P} & 0 \\ * & * & * & * & -\bar{P} \end{bmatrix} \\ &= \begin{bmatrix} \tilde{\Sigma}_{11} & 0 & \bar{C}_i^T K^T & \bar{A}_{ij}^T \bar{P} & \bar{E}_i^T \bar{P} \\ * & -H^TQH & 0 & H^T \bar{A}_{dij}^T \bar{P} & H^T \bar{E}_{di}^T \bar{P} \\ * & * & -2I & \bar{B}_{fj}^T \bar{P} & 0 \\ * & * & * & -\bar{P} & 0 \\ * & * & * & * & -\bar{P} \end{bmatrix} + \tilde{M}_{1i} F_{1i} \tilde{N}_{1i} + \tilde{N}_{1i}^T F_{1i} \tilde{M}_{1i}^T \\ &+ \tilde{M}_{2i} F_{2i} \tilde{N}_{2i} + \tilde{N}_{2i}^T F_{2i} \tilde{M}_{2i}^T.\end{aligned}\quad (3.10)$$

By Lemmas 2.2 and 2.3, from (3.8)–(3.10), we have

$$\Gamma = \begin{bmatrix} \Gamma_{11} & \Gamma_{12} & \Gamma_{13} \\ * & \Gamma_{22} & \Gamma_{23} \\ * & * & \Gamma_{33} \end{bmatrix} < 0, \quad (3.11)$$

where

$$\begin{aligned}\Gamma_{11} &= \tilde{\Sigma}_{11} + (\bar{A}_{ij} + \Delta \bar{A}_i(k))^T \bar{P} (\bar{A}_{ij} + \Delta \bar{A}_i(k)) + (\bar{E}_i + \Delta \bar{E}_i(k))^T \bar{P} (\bar{E}_i + \Delta \bar{E}_i(k)), \\ \Gamma_{12} &= (\bar{A}_{ij} + \Delta \bar{A}_i(k))^T \bar{P} (\bar{A}_{dij} + \Delta \bar{A}_{dij}(k)) H + (\bar{E}_i + \Delta \bar{E}_i(k))^T \bar{P} (\bar{E}_{di} + \Delta \bar{E}_{di}(k)) H, \\ \Gamma_{13} &= \bar{C}_i^T K^T + (\bar{A}_{ij} + \Delta \bar{A}_i(k))^T \bar{P} \bar{B}_{fj}, \\ \Gamma_{22} &= H^T \left[-Q + (\bar{A}_{dij} + \Delta \bar{A}_{dij}(k))^T \bar{P} (\bar{A}_{dij} + \Delta \bar{A}_{dij}(k)) \right. \\ &\quad \left. + (\bar{E}_{di} + \Delta \bar{E}_{di}(k))^T \bar{P} (\bar{E}_{di} + \Delta \bar{E}_{di}(k)) \right] H,\end{aligned}$$

$$\begin{aligned}\Gamma_{23} &= H^T \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right)^T \bar{P} \bar{B}_{fj}, \\ \Gamma_{33} &= -2I + \bar{B}_{fj}^T \bar{P} \bar{B}_{fj}.\end{aligned}\tag{3.12}$$

Then there exists a small scalar $\alpha > 0$ such that

$$\Gamma < \begin{bmatrix} -\alpha I & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}.\tag{3.13}$$

Consider the Lyapunov-Krasovskii functional candidate as follows:

$$V(k) = V_1(k) + V_2(k) + V_3(k),\tag{3.14}$$

where

$$\begin{aligned}V_1(k) &= e^T(k) \bar{P} e(k), \\ V_2(k) &= \sum_{i=k-\tau(k)}^{k-1} e^T(i) H^T Q H e(i), \\ V_3(k) &= \sum_{j=-\tau_2+2}^{-\tau_1+1} \sum_{l=k+j-1}^{k-1} e^T(l) H^T Q H e(l).\end{aligned}\tag{3.15}$$

Calculating the difference of $V(k)$ along the filtering error system (2.14) with $v(k) = 0$ and taking the mathematical expectation, we have

$$\begin{aligned}\mathbf{E}\{\Delta V(k)\} &= \mathbf{E}\{\Delta V_1(k)\} + \mathbf{E}\{\Delta V_2(k)\} + \mathbf{E}\{\Delta V_3(k)\}, \\ \mathbf{E}\{\Delta V_1(k)\} &= \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k)) h_j(\theta(k)) \mathbf{E}\left\{ \left\{ \left(\bar{A}_{ij} + \Delta \bar{A}_i(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) \right. \right. \\ &\quad \left. \left. + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) H e(k - \tau(k)) \right. \right. \\ &\quad \left. \left. + \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) H e(k - \tau(k)) \right] \right. \right. \\ &\quad \left. \left. \times w(k) \right\}^T \bar{P} \times \left\{ \left(\bar{A}_{ij} + \Delta \bar{A}_i(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) \right. \right. \\ &\quad \left. \left. + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) H e(k - \tau(k)) \right. \right. \\ &\quad \left. \left. + \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) \right. \right. \\ &\quad \left. \left. \times H e(k - \tau(k)) \right] \times w(k) \right\} - e^T(k) \bar{P} e(k) \Big\}.\end{aligned}\tag{3.16}$$

Noting (2.2) and (2.12), we have

$$\begin{aligned}
\mathbf{E}\{\Delta V_1(k)\} &\leq \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k))h_j(\theta(k)) \left\{ \left[\left(\bar{A}_{ij} + \Delta \bar{A}_i(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) \right. \right. \\
&\quad \times \left. \left. H e(k - \tau(k)) \right] \bar{P} \left[\left(\bar{A}_{ij} + \Delta \bar{A}_i(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) \right. \right. \\
&\quad \left. \left. + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) H e(k - \tau(k)) \right] \right. \\
&\quad \left. + \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) H e(k - \tau(k)) \right] \bar{P} \right. \\
&\quad \left. \times \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) H e(k - \tau(k)) \right] \right. \\
&\quad \left. - e(k)^T \bar{P} e(k) - 2\phi_s \left(\bar{C}_i e(k) \right) \left[\phi_s \left(\bar{C}_i e(k) \right) - K \bar{C}_i e(k) \right] \right\}. \tag{3.17}
\end{aligned}$$

By some calculations, we have

$$\begin{aligned}
\mathbf{E}\{\Delta V_2(k)\} &= \sum_{i=k+1-\tau(k+1)}^{k-\tau_1} e^T(i) H^T Q H e(i) - e^T(k - \tau(k)) H^T Q H e(k - \tau(k)) \\
&\quad + e^T(k) H^T Q H e(k) + \sum_{i=k+1-\tau_1}^{k-1} e^T(i) H^T Q H e(i) - \sum_{i=k+1-\tau(k)}^{k-1} e^T(i) H^T Q H e(i), \tag{3.18}
\end{aligned}$$

$$\begin{aligned}
\mathbf{E}\{\Delta V_3(k)\} &= \sum_{j=-\tau_2+2}^{-\tau_1+1} \left[e^T(k) H^T Q H e(k) + \sum_{l=k+j}^{k-1} e^T(l) H^T Q H e(l) - \sum_{l=k+j-1}^{k-1} e^T(l) H^T Q H e(l) \right] \\
&= (\tau_2 - \tau_1) e^T(k) H^T Q H e(k) - \sum_{i=k+1-\tau_2}^{k-\tau_1} e^T(i) H^T Q H e(i). \tag{3.19}
\end{aligned}$$

Since $\tau_1 \leq \tau(k) \leq \tau_2$, we have

$$\sum_{i=k+1-\tau_1}^{k-1} e^T(i) H^T Q H e(i) - \sum_{i=k+1-\tau(k)}^{k-1} e^T(i) H^T Q H e(i) \leq 0, \tag{3.20}$$

$$\sum_{i=k+1-\tau(k+1)}^{k-\tau_1} e^T(i) H^T Q H e(i) - \sum_{i=k+1-\tau_2}^{k-\tau_1} e^T(i) H^T Q H e(i) \leq 0. \tag{3.21}$$

Combining (3.18)–(3.21), we have

$$\mathbf{E}\{\Delta V_2(k) + \Delta V_3(k)\} \leq (\tau_2 - \tau_1 + 1)e^T(k)H^T Q H e(k) - e^T(k - \tau(k))H^T Q H e(k - \tau(k)). \quad (3.22)$$

Combining (3.17) and (3.22) yields

$$\begin{aligned} & \mathbf{E}\{\Delta V(k)\} \\ & \leq \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k))h_j(\theta(k)) \left\{ \left[\left(\bar{A}_{ij} + \Delta \bar{A}_i(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) \right. \right. \\ & \quad \times H e(k - \tau(k)) \left. \right]^T \bar{P} \left[\left(\bar{A}_{ij} + \Delta \bar{A}_i(k) \right) e(k) + \bar{B}_{fj} \phi_s \left(\bar{C}_i e(k) \right) \right. \\ & \quad \left. \left. + \left(\bar{A}_{dij} + \Delta \bar{A}_{dij}(k) \right) H e(k - \tau(k)) \right] \right. \\ & \quad + \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) H e(k - \tau(k)) \right]^T \bar{P} \\ & \quad \times \left[\left(\bar{E}_i + \Delta \bar{E}_i(k) \right) e(k) + \left(\bar{E}_{di} + \Delta \bar{E}_{di}(k) \right) H e(k - \tau(k)) \right] \\ & \quad + e^T(k) \left[(\tau_2 - \tau_1 + 1) H^T Q H - \bar{P} \right] e(k) - 2\phi_s \left(\bar{C}_i e(k) \right) \\ & \quad \times \left[\phi_s \left(\bar{C}_i e(k) \right) - K \bar{C}_i e(k) \right] \\ & \quad \left. - e^T(k - \tau(k)) H^T Q H e(k - \tau(k)) \right\} \\ & = \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k))h_j(\theta(k)) \zeta^T(k) \Gamma \zeta(k), \end{aligned} \quad (3.23)$$

where

$$\zeta^T(k) = \left[e^T(k) \quad e^T(k - \tau(k)) \quad \phi_s^T \left(\bar{C}_i e(k) \right) \right], \quad (3.24)$$

and Γ is given in (3.11). Thus, it follows from (3.13) and (3.23) that

$$\mathbf{E}\{V(k+1)\} - \mathbf{E}\{V(k)\} < -\alpha \mathbf{E}\{|e(k)|^2\}. \quad (3.25)$$

Hence, by summing up both sides of (3.25) from 0 to N for any integer $N > 1$, we have

$$\mathbf{E}\{V(N+1)\} - \mathbf{E}\{V(0)\} < -\alpha \mathbf{E}\left\{ \sum_{k=0}^N |e(k)|^2 \right\}, \quad (3.26)$$

which yields

$$\mathbf{E} \left\{ \sum_{k=0}^N |e(k)|^2 \right\} < \frac{1}{\alpha} [\mathbf{E}\{V(0)\} - \mathbf{E}\{V(N+1)\}] \leq \frac{1}{\alpha} \mathbf{E}\{V(0)\} \leq c \mathbf{E}\{|x(0)|^2\}, \quad (3.27)$$

where $c = (1/\alpha)\lambda_{\max}(\bar{P})$. Taking $N \rightarrow \infty$, it is shown from (2.17) and (3.27) that the filtering error system (2.14) is robustly stochastic stable for $v(k) = 0$.

Next, we will show that the filtering error system (2.14)-(2.15) satisfies

$$\|\tilde{z}(k)\|_{e_2} < \gamma \|v(k)\|_{e_2} \quad (3.28)$$

for all nonzero $v(k) \in L_{e_2}([0, \infty), \mathbb{R}^p)$. To this end, define

$$J(N) = \mathbf{E} \left\{ \sum_{k=1}^N [|\tilde{z}(k)|^2 - \gamma^2 |v(k)|^2] \right\} \quad (3.29)$$

with any integer $N > 0$. Then for any nonzero $v(k)$, we have

$$\begin{aligned} J(N) &= \mathbf{E} \left\{ \sum_{k=1}^N |\tilde{z}(k)|^2 - \gamma^2 |v(k)|^2 \right\} \\ &\leq \mathbf{E} \left\{ \sum_{k=1}^N [|\tilde{z}(k)|^2 - \gamma^2 |v(k)|^2 + \Delta V(k)] \right\} - \mathbf{E}\{V(N+1)\} \\ &= \mathbf{E} \left\{ \sum_{k=1}^N [|\tilde{z}(k)|^2 - \gamma^2 |v(k)|^2 + \Delta V(k)] \right\} \\ &= \mathbf{E} \left\{ \sum_{k=1}^N \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k)) h_j(\theta(k)) \tilde{\xi}^T(k) \hat{\Gamma}_{ij} \tilde{\xi}(k) \right\}, \end{aligned} \quad (3.30)$$

where

$$\begin{aligned} \tilde{\xi}^T(k) &= \left[e^T(k) \quad e^T(k - \tau(k)) \quad v^T(k) \quad \phi_s(\bar{C}_i e(k))^T \right], \\ \hat{\Gamma}_{ij} &= \begin{bmatrix} \hat{\Gamma}_{11} & \hat{\Gamma}_{12} & \hat{\Gamma}_{13} & \hat{\Gamma}_{14} \\ * & \hat{\Gamma}_{22} & \hat{\Gamma}_{23} & \hat{\Gamma}_{24} \\ * & * & \hat{\Gamma}_{33} & \hat{\Gamma}_{34} \\ * & * & * & \hat{\Gamma}_{44} \end{bmatrix}, \end{aligned}$$

$$\begin{aligned}
\hat{\Gamma}_{11} &= \tilde{\Sigma}_{11} + \left(\bar{A}_{ij} + \Delta\bar{A}_i(k)\right)^T \bar{P} \left(\bar{A}_{ij} + \Delta\bar{A}_i(k)\right) \\
&\quad + \left(\bar{E}_i + \Delta\bar{E}_i(k)\right)^T \bar{P} \left(\bar{E}_i + \Delta\bar{E}_i(k)\right) + (L_i - L_j)^T (L_i - L_j), \\
\hat{\Gamma}_{12} &= \left(\bar{A}_{ij} + \Delta\bar{A}_i(k)\right)^T \bar{P} \left(\bar{A}_{dij} + \Delta\bar{A}_{dij}(k)\right) H + \left(\bar{E}_i + \Delta\bar{E}_i(k)\right)^T \bar{P} \left(\bar{E}_{di} + \Delta\bar{E}_{di}(k)\right) H, \\
\hat{\Gamma}_{13} &= \left(\bar{A}_{ij} + \Delta\bar{A}_i(k)\right)^T \bar{P} \left(\bar{B}_{ij} + \Delta\bar{B}_{ij}(k)\right) + \left(\bar{E}_i + \Delta\bar{E}_i(k)\right)^T \bar{P} \left(\bar{G}_i + \Delta\bar{G}_i(k)\right), \\
\hat{\Gamma}_{14} &= \bar{C}_i^T K^T + \left(\bar{A}_{ij} + \Delta\bar{A}_i(k)\right)^T \bar{P} \bar{B}_{fj}, \\
\hat{\Gamma}_{22} &= H^T \left[-Q + \left(\bar{A}_{dij} + \Delta\bar{A}_{dij}(k)\right)^T \bar{P} \left(\bar{A}_{dij} + \Delta\bar{A}_{dij}(k)\right) \right. \\
&\quad \left. + \left(\bar{E}_{di} + \Delta\bar{E}_{di}(k)\right)^T \bar{P} \left(\bar{E}_{di} + \Delta\bar{E}_{di}(k)\right) \right] H, \\
\hat{\Gamma}_{23} &= H^T \left(\bar{A}_{dij} + \Delta\bar{A}_{dij}(k)\right)^T \bar{P} \left(\bar{B}_{ij} + \Delta\bar{B}_{ij}(k)\right) + H^T \left(\bar{E}_{di} + \Delta\bar{E}_{di}(k)\right)^T \bar{P} \left(\bar{G}_i + \Delta\bar{G}_i(k)\right), \\
\hat{\Gamma}_{24} &= H^T \left(\bar{A}_{dij} + \Delta\bar{A}_{dij}(k)\right)^T \bar{P} \bar{B}_{fj}, \\
\hat{\Gamma}_{33} &= -\gamma^2 I + \left(\bar{B}_{ij} + \Delta\bar{B}_{ij}(k)\right)^T \bar{P} \left(\bar{B}_{ij} + \Delta\bar{B}_{ij}(k)\right) + \left(\bar{G}_i + \Delta\bar{G}_i(k)\right)^T \bar{P} \left(\bar{G}_i + \Delta\bar{G}_i(k)\right), \\
\hat{\Gamma}_{34} &= \left(\bar{B}_{ij} + \Delta\bar{B}_{ij}(k)\right)^T \bar{P} \bar{B}_{fj}, \\
\hat{\Gamma}_{44} &= -2I + \bar{B}_{fj}^T \bar{P} \bar{B}_{fj}.
\end{aligned} \tag{3.31}$$

It can be shown that if there exist real matrices $X > 0$, $Y > 0$, $Q > 0$, any matrices A_{Fj}, B_{Fj} , scalars $\varepsilon_{1i} > 0$, $\varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$ satisfying LMI (3.1). Since $v(k) \in L_{e2}([0, \infty), \mathbb{R}^p) \neq 0$ and by Lemma 2.2, it is implied that $\hat{\Gamma}_{ij} < 0$, and, thus, $J(N) < 0$. That is, $\|\tilde{z}(k)\|_{e2} < \gamma \|v(k)\|_{e2}$. This completes the proof. \square

Remark 3.2. When τ_1 and τ_2 are given, matrix inequality (3.1) is linear matrix inequalities in matrix variables $X > 0$, $Y > 0$, $Q > 0$, A_{Fj}, B_{Fj} , scalars $\varepsilon_{1i} > 0$, $\varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$, which can be efficiently solved by the developed interior point algorithm [33]. Meanwhile, it is easy to find the minimal attenuation level γ .

Remark 3.3. Theorem 3.1 is suggested to the H_∞ filter design for uncertain discrete-time fuzzy stochastic systems with sensor nonlinearities and time-varying delay. This approach is called fuzzy-rule-dependent approach, which is applicable to the case that the information of the premise variable $\theta_i(k)$ ($i = 1, 2, \dots, r$) is available; for example, see [24]. For the special case that the premise variables are unavailable, the fuzzy-rule-independent approach as in [24, 29] is adopted, which may result in larger conservativeness due to lack of the information of the premise variables.

In the following, we will present the fuzzy-rule-dependent H_∞ filter design for uncertain discrete-time fuzzy stochastic system with time-varying delay, which are less conservative than Theorem 3.1.

Theorem 3.4. Consider the uncertain discrete-time fuzzy stochastic systems in (2.1). A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system (2.14)-(2.15) is robustly stochastic stable with performance γ , if there exist real matrices $X > 0$, $Y > 0$, $Q > 0$, any matrices A_{Fi} , B_{Fi} , scalars $\varepsilon_{1i} > 0$, $\varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$, $i < j$, such that the following LMIs are satisfied:

$$\Sigma_{ii} < 0, \quad i = 1, 2, \dots, r, \quad (3.32)$$

$$\Sigma_{ij} + \Sigma_{ji} < 0, \quad i, j = 1, 2, \dots, r, \quad i < j, \quad (3.33)$$

where

$$\Sigma_{ij} = \begin{bmatrix} \Sigma_{11} & 0 & 0 & 0 & C_i^T K^T & A_i^T X & \Sigma_{17} & E_i^T X & E_i^T Y & 0 & 0 & 0 \\ * & \Sigma_{22} & 0 & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 & 0 \\ * & * & \Sigma_{33} & 0 & 0 & A_{di}^T X & \Sigma_{37} & E_{di}^T X & E_{di}^T Y & 0 & 0 & 0 \\ * & * & * & \Sigma_{44} & 0 & B_i^T X & \Sigma_{47} & G_i^T X & G_i^T Y & 0 & 0 & 0 \\ * & * & * & * & -2I & 0 & -B_{Fj}^T & 0 & 0 & 0 & 0 & 0 \\ * & * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{3i} & XM_{6i} \\ * & * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Sigma_{711} & \Sigma_{712} \\ * & * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{4i} & XM_{7i} \\ * & * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{4i} & YM_{7i} \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{1i}I & 0 & 0 \\ * & * & * & * & * & * & * & * & * & * & -\varepsilon_{2i}I & 0 \\ * & * & * & * & * & * & * & * & * & * & * & -\varepsilon_{3i}I \end{bmatrix}, \quad (3.34)$$

where Σ_{11} , Σ_{22} , Σ_{33} , Σ_{44} , Σ_{17} , Σ_{37} , Σ_{47} , Σ_{711} , and Σ_{712} are defined in Theorem 3.1. Moreover, if the previous conditions are satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Proof. Considering (2.2) and (3.1), it follows from (3.30) that

$$\begin{aligned} J(N) &\leq \mathbf{E} \left\{ \sum_{k=1}^N \sum_{i=1}^r \sum_{j=1}^r h_i(\theta(k)) h_j(\theta(k)) \tilde{\zeta}^T(k) \hat{\Gamma}_{ij} \hat{\zeta}(k) \right\} \\ &= \mathbf{E} \left\{ \sum_{k=1}^N \tilde{\zeta}^T(k) \left[\sum_{i=1}^r h_i^2(\theta(k)) \hat{\Gamma}_{ii} + \sum_{i=1}^r \sum_{i < j}^r h_i(\theta(k)) h_j(\theta(k)) (\hat{\Gamma}_{ij} + \hat{\Gamma}_{ji}) \right] \hat{\zeta}(k) \right\}. \end{aligned} \quad (3.35)$$

Our elaborate estimation $J(N)$ is negative definite if and only if $\hat{\Gamma}_{ii} < 0$ and $\hat{\Gamma}_{ij} + \hat{\Gamma}_{ji} < 0$. By Lemmas 2.2 and 2.3, we can obtain that (3.32) is equivalent to $\hat{\Gamma}_{ii} < 0$, and (3.33) is equivalent to $\hat{\Gamma}_{ij} + \hat{\Gamma}_{ji} < 0$. Thus, we can conclude that the LMIs (3.32) and (3.33) can guarantee $J(N) < 0$. That is, $\|\tilde{z}(k)\|_{e2} < \gamma \|v(k)\|_{e2}$. This completes the proof. \square

Theorem 3.5. Consider the uncertain discrete-time fuzzy stochastic systems in (2.1). A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system (2.14)-(2.15) is robustly stochastic stable with performance γ , if there exist real matrices $X > 0$, $Y > 0$, $Q > 0$, any matrices A_{Fi} , B_{Fi} , scalars $\varepsilon_{1i} > 0$, $\varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$, $i < j$, such that the following LMIs are satisfied:

$$\Sigma_{ii} < \Omega_{ii}, \quad i = 1, 2, \dots, r, \quad (3.36)$$

$$\Sigma_{ij} + \Sigma_{ji} < \Omega_{ij} + \Omega_{ij}^T, \quad i, j = 1, 2, \dots, r, \quad i < j, \quad (3.37)$$

$$\Omega = \begin{bmatrix} \Omega_{11} & \Omega_{12} & \cdots & \Omega_{1r} \\ * & \Omega_{22} & \cdots & \Omega_{2r} \\ \vdots & \vdots & \ddots & \vdots \\ * & * & \cdots & \Omega_{rr} \end{bmatrix} < 0, \quad (3.38)$$

where Σ_{ij} , Σ_{11} , Σ_{22} , Σ_{33} , Σ_{44} , Σ_{17} , Σ_{37} , Σ_{47} , Σ_{711} , and Σ_{712} are defined in Theorems 3.4 and 3.1, respectively. Moreover, if the previous conditions are satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Proof. This theorem can be proved by employing the same techniques as in the proof of Theorem 3.4; hence, the detailed procedure is omitted here. \square

Remark 3.6. It is noted that the convexifying procedure proposed in this paper is based on the relaxation inequality (3.36), and extensions of the current derivations based on the more powerful relaxation techniques presented in [8, 24, 30] are straightforward. In this way, the design conservatism can be further reduced but the computation complexity will also increase. By considering the information on the premises [8, 24, 30], Theorems 3.4 and 3.5 relaxed the conservatism of the previous works [29] by representing the interactions among the fuzzy subsystems in a matrix and solving it in a numerical manner.

For comparison, if there is no time-varying delay exit in (2.1), we consider the following discrete-time stochastic systems [29]:

$$\begin{aligned} x(k+1) &= (A_i + \Delta A_i(k))x(k) + (B_i + \Delta B_i(k))v(k) \\ &\quad + [(E_i + \Delta E_i(k))x(k) + (G_i + \Delta G_i(k))v(k)]w(k), \\ y(k) &= \phi(C_i(k))x(k) + (D_i + \Delta D_i(k))v(k), \\ z(k) &= L_i x(k), \quad i = 1, 2, \dots, r. \end{aligned} \quad (3.39)$$

Remark 3.7. This class of uncertain discrete-time fuzzy stochastic systems with sensor nonlinearities has been considered in [29]; a new different Lyapunov functional is then employed to deal with systems with sensor nonlinearities. Then from Theorems 3.1, 3.4, and 3.5, we can get the following corollaries, which are less conservative than Theorem 3.1 in [29].

Corollary 3.8. Consider the uncertain discrete-time fuzzy stochastic systems in (3.39). A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system is robustly stochastic stable with

performance γ , if there exist real matrices $X > 0, Y > 0$, any matrices A_{Fj}, B_{Fj} , scalars $\varepsilon_{1i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$ such that the following LMI is satisfied:

$$\Pi = \begin{bmatrix} \Pi_{11} & 0 & 0 & C_i^T K^T & A_i^T X & \Pi_{16} & E_i^T X & E_i^T Y & 0 & 0 \\ * & \Pi_{22} & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 \\ * & * & \Pi_{33} & 0 & B_i^T X & \Pi_{36} & G_i^T X & G_i^T Y & 0 & 0 \\ * & * & * & -2I & 0 & -B_{Fj}^T & 0 & 0 & 0 & 0 \\ * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{6i} \\ * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Pi_{610} \\ * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{7i} \\ * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{7i} \\ * & * & * & * & * & * & * & * & -\varepsilon_{1i} I & 0 \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{3i} I \end{bmatrix} < 0, \quad (3.40)$$

where

$$\begin{aligned} \Pi_{11} &= -X + \varepsilon_{1i} N_{1i}^T N_{1i}, & \Pi_{22} &= -Y + (L_i - L_j)^T (L_i - L_j), & \Pi_{33} &= -\gamma^2 I + \varepsilon_{3i} N_{3i}^T N_{3i}, \\ \Pi_{16} &= A_i^T Y - A_{Fj}^T - C_i^T K_1^T B_{Fj}^T, & \Pi_{36} &= B_i^T Y - D_i^T B_{Fj}^T, & \Pi_{610} &= YM_{6i} - B_{Fj} M_{8i}. \end{aligned} \quad (3.41)$$

Moreover, if the previous condition is satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Corollary 3.9. Consider the uncertain discrete-time fuzzy stochastic systems in (3.39). Give a filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system is robustly stochastic stable with performance γ , if there exist real matrices $X > 0, Y > 0$, any matrices A_{Fj}, B_{Fj} , scalars $\varepsilon_{1i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$ such that the following LMIs are satisfied:

$$\begin{aligned} \Pi_{ii} &< 0, \quad i = 1, 2, \dots, r, \\ \Pi_{ij} + \Pi_{ji} &< 0, \quad i, j = 1, 2, \dots, r, \quad i < j, \end{aligned}$$

$$\Pi_{ij} = \begin{bmatrix} \Pi_{11} & 0 & 0 & C_i^T K^T & A_i^T X & \Pi_{16} & E_i^T X & E_i^T Y & 0 & 0 \\ * & \Pi_{22} & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 \\ * & * & \Pi_{33} & 0 & B_i^T X & \Pi_{36} & G_i^T X & G_i^T Y & 0 & 0 \\ * & * & * & -2I & 0 & -B_{Fj}^T & 0 & 0 & 0 & 0 \\ * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{6i} \\ * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Pi_{610} \\ * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{7i} \\ * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{7i} \\ * & * & * & * & * & * & * & * & -\varepsilon_{1i} I & 0 \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{3i} I \end{bmatrix}, \quad (3.42)$$

where $\Pi_{11}, \Pi_{22}, \Pi_{33}, \Pi_{16}, \Pi_{36}, \Pi_{610}$ are defined in Corollary 3.8. Moreover, if the previous conditions are satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Corollary 3.10. Consider the uncertain discrete-time fuzzy stochastic systems in (3.39). A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system is robustly stochastic stable with performance γ , if there exist real matrices $X > 0, Y > 0$, any matrices A_{Fj}, B_{Fj} , scalars $\varepsilon_{1i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$ such that the following LMIs are satisfied:

$$\begin{aligned} \Pi_{ii} &< \Omega_{ii}, \quad i = 1, 2, \dots, r, \\ \Pi_{ij} + \Pi_{ji} &< \Omega_{ij} + \Omega_{ji}^T, \quad i, j = 1, 2, \dots, r, \quad i < j, \\ \Omega_{ij} &= \begin{bmatrix} \Omega_{11} & \Omega_{12} & \cdots & \Omega_{1r} \\ * & \Omega_{22} & \cdots & \Omega_{2r} \\ \vdots & \vdots & \ddots & \vdots \\ * & * & \cdots & \Omega_{rr} \end{bmatrix} < 0, \end{aligned} \quad (3.43)$$

where $\Pi_{11}, \Pi_{22}, \Pi_{33}, \Pi_{16}, \Pi_{36}, \Pi_{610}$ are defined in Corollary 3.8. Moreover, if the previous conditions are satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

In the sequel, special results for the case of linear sensor, that is, $\phi(Cx(k)) = Cx(k)$, may be obtained from Theorems 3.1, 3.4, and 3.5.

Corollary 3.11. Consider the uncertain discrete-time fuzzy stochastic systems in (2.1) with $\phi(Cx(k)) = Cx(k)$. A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system is robustly stochastic stable with performance γ , if there exist real matrices $X > 0, Y > 0, Q > 0$, any matrices A_{Fj}, B_{Fj} , scalars $\varepsilon_{1i} > 0, \varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r$ such that the following LMI is satisfied:

$$\begin{bmatrix} \Sigma_{11} & 0 & 0 & 0 & A_i^T X & \bar{\Sigma}_{17} & E_i^T X & E_i^T Y & 0 & 0 & 0 \\ * & \Sigma_{22} & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 & 0 \\ * & * & \Sigma_{33} & 0 & A_{di}^T X & \Sigma_{37} & E_{di}^T X & E_{di}^T Y & 0 & 0 & 0 \\ * & * & * & \Sigma_{44} & B_i^T X & \Sigma_{47} & G_i^T X & G_i^T Y & 0 & 0 & 0 \\ * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{3i} & XM_{6i} \\ * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Sigma_{711} & \Sigma_{712} \\ * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{4i} & XM_{7i} \\ * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{4i} & YM_{7i} \\ * & * & * & * & * & * & * & * & -\varepsilon_{1i} I & 0 & 0 \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{2i} I & 0 \\ * & * & * & * & * & * & * & * & * & * & -\varepsilon_{3i} I \end{bmatrix} < 0, \quad (3.44)$$

where

$$\bar{\Sigma}_{17} = A_i^T Y - A_{Fj}^T - C_i^T B_{Fj}^T. \quad (3.45)$$

Moreover, if the previous condition is satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Corollary 3.12. Consider the uncertain discrete-time fuzzy stochastic systems in (2.1) with $\phi(Cx(k)) = Cx(k)$. A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system is robustly stochastic stable with performance γ , if there exist real matrices $X > 0, Y > 0, Q > 0$, any matrices A_{Fi}, B_{Fi} , scalars $\varepsilon_{1i} > 0, \varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r, i < j$, such that the following LMIs are satisfied:

$$\begin{aligned} \bar{\Sigma}_{ii} &< 0, \quad i = 1, 2, \dots, r, \\ \bar{\Sigma}_{ij} + \bar{\Sigma}_{ji} &< 0, \quad i, j = 1, 2, \dots, r, \quad i < j, \end{aligned} \tag{3.46}$$

where

$$\bar{\Sigma}_{ij} = \begin{bmatrix} \Sigma_{11} & 0 & 0 & 0 & A_i^T X & \bar{\Sigma}_{17} & E_i^T X & E_i^T Y & 0 & 0 & 0 \\ * & \Sigma_{22} & 0 & 0 & 0 & A_{Fj}^T & 0 & 0 & 0 & 0 & 0 \\ * & * & \Sigma_{33} & 0 & A_{di}^T X & \Sigma_{37} & E_{di}^T X & E_{di}^T Y & 0 & 0 & 0 \\ * & * & * & \Sigma_{44} & B_i^T X & \Sigma_{47} & G_i^T X & G_i^T Y & 0 & 0 & 0 \\ * & * & * & * & -X & 0 & 0 & 0 & XM_{1i} & XM_{3i} & XM_{6i} \\ * & * & * & * & * & -Y & 0 & 0 & YM_{1i} & \Sigma_{711} & \Sigma_{712} \\ * & * & * & * & * & * & -X & 0 & XM_{2i} & XM_{4i} & XM_{7i} \\ * & * & * & * & * & * & * & -Y & YM_{2i} & YM_{4i} & YM_{7i} \\ * & * & * & * & * & * & * & * & -\varepsilon_{1i}I & 0 & 0 \\ * & * & * & * & * & * & * & * & * & -\varepsilon_{2i}I & 0 \\ * & * & * & * & * & * & * & * & * & * & -\varepsilon_{3i}I \end{bmatrix}, \tag{3.47}$$

where $\Sigma_{11}, \Sigma_{22}, \Sigma_{33}, \Sigma_{44}, \Sigma_{37}, \Sigma_{47}, \Sigma_{711}, \Sigma_{712}$, and $\bar{\Sigma}_{17}$ are defined in Theorem 3.1 and Corollary 3.11, respectively. Moreover, if the previous conditions are satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Corollary 3.13. Consider the uncertain discrete-time fuzzy stochastic systems in (2.9) with $\phi(Cx(k)) = Cx(k)$. A filter of form (2.13) and constants τ_1 and τ_2 , the filtering error system is robustly stochastic stable with performance γ , if there exist real matrices $X > 0, Y > 0, Q > 0$, any matrices A_{Fi}, B_{Fi} , scalars $\varepsilon_{1i} > 0, \varepsilon_{2i} > 0$ and $\varepsilon_{3i} > 0$ for $i, j = 1, 2, \dots, r, i < j$, such that the following LMIs are satisfied:

$$\begin{aligned} \bar{\Sigma}_{ii} &< \bar{\Omega}_{ii}, \quad i = 1, 2, \dots, r, \\ \bar{\Sigma}_{ij} + \bar{\Sigma}_{ji} &< \bar{\Omega}_{ij} + \bar{\Omega}_{ij}^T, \quad i, j = 1, 2, \dots, r, \quad i < j, \end{aligned}$$

$$\bar{\Omega} = \begin{bmatrix} \bar{\Omega}_{11} & \bar{\Omega}_{12} & \cdots & \bar{\Omega}_{1r} \\ * & \bar{\Omega}_{22} & \cdots & \bar{\Omega}_{2r} \\ \vdots & \vdots & \ddots & \vdots \\ * & * & \cdots & \bar{\Omega}_{rr} \end{bmatrix} < 0, \tag{3.48}$$

Table 1: Minimum index γ for different methods.

| Methods | Theorem 3.1 [29] | Corollary 3.8 | Corollary 3.9 | Corollary 3.10 |
|-----------------|------------------|---------------|---------------|----------------|
| γ_{\min} | 0.2931 | 0.2030 | 0.1869 | 0.1376 |

where $\bar{\Sigma}_{ij}, \Sigma_{11}, \Sigma_{22}, \Sigma_{33}, \Sigma_{44}, \Sigma_{37}, \Sigma_{47}, \Sigma_{711}, \Sigma_{712}$, and $\bar{\Sigma}_{17}$ are defined in Corollary 3.12, Theorem 3.1, and Corollary 3.11, respectively. Moreover, if the previous conditions are satisfied, an acceptable state-space realization of the H_∞ filter is given by LMI (3.3).

Remark 3.14. In the previous results, there still exists conservativeness; for these, we can also use the delay-dependent stability results of discrete-time systems in [13–16] to derive the less conservative theorems and corollaries for filtering problem of uncertain discrete-time fuzzy stochastic systems with sensor nonlinearities and time-varying delay.

4. Numerical Example

In this section, two simulation examples are given to illustrate the effectiveness and benefits of the proposed approach.

Example 4.1. Consider the system (3.39) with parameters as follows:

$$\begin{aligned}
A_1 &= \begin{bmatrix} -0.2 & 0.1 \\ 0.1 & -0.3 \end{bmatrix}, & B_1 &= \begin{bmatrix} 0.1 \\ 0.2 \end{bmatrix}, & E_1 &= \begin{bmatrix} -0.2 & 0.3 \\ 0.1 & 0.3 \end{bmatrix}, & G_1 &= \begin{bmatrix} -0.1 \\ 0.3 \end{bmatrix}, \\
C_1 &= \begin{bmatrix} -0.2 & 0.3 \\ 0.1 & 0.4 \end{bmatrix}, & D_1 &= \begin{bmatrix} 0.2 \\ 0.1 \end{bmatrix}, & L_1 &= \begin{bmatrix} 0.3 & 0.2 \\ 0.2 & 0.4 \end{bmatrix}, \\
M_{11} &= \begin{bmatrix} 0.1 \\ 0.1 \end{bmatrix}, & M_{21} &= \begin{bmatrix} -0.4 \\ -0.5 \end{bmatrix}, & M_{61} &= \begin{bmatrix} -0.2 \\ 0.1 \end{bmatrix}, & M_{71} &= \begin{bmatrix} -0.2 \\ 0.3 \end{bmatrix}, \\
M_{81} &= \begin{bmatrix} -0.2 \\ 0.3 \end{bmatrix}, & N_{11} &= [0.2 \ 0.1], & N_{31} &= 0.2, & K_1 &= \text{diag}\{0.4, 0.6\}, \\
A_2 &= \begin{bmatrix} -0.1 & -0.2 \\ 0.2 & 0.3 \end{bmatrix}, & B_2 &= \begin{bmatrix} 0.2 \\ -0.2 \end{bmatrix}, & E_2 &= \begin{bmatrix} -0.2 & -0.2 \\ 0.3 & 0.1 \end{bmatrix}, & G_2 &= \begin{bmatrix} -0.2 \\ 0.2 \end{bmatrix}, \\
C_2 &= \begin{bmatrix} -0.1 & -0.2 \\ 0.3 & 0.3 \end{bmatrix}, & D_2 &= \begin{bmatrix} 0.4 \\ 0.3 \end{bmatrix}, & L_2 &= \begin{bmatrix} -0.1 & 0.3 \\ 0.2 & 0.3 \end{bmatrix}, \\
M_{12} &= \begin{bmatrix} -0.2 \\ 0.2 \end{bmatrix}, & M_{22} &= \begin{bmatrix} 0.3 \\ 0.1 \end{bmatrix}, & M_{62} &= \begin{bmatrix} 0.3 \\ 0.1 \end{bmatrix}, & M_{72} &= \begin{bmatrix} 0.1 \\ 0.3 \end{bmatrix}, \\
M_{82} &= \begin{bmatrix} 0.1 \\ -0.2 \end{bmatrix}, & N_{12} &= [0.2 \ 0.3], & N_{32} &= 0.2, & K_2 &= \text{diag}\{0.1, 0.8\}.
\end{aligned} \tag{4.1}$$

This system has been considered in [29]. For comparison with the Theorem 3.1 in [29], the computation results of γ_{\min} under Corollaries 3.8, 3.9, and 3.10 are listed in Table 1.

This example conclusively shows that our results are less conservative than Theorem 3.1 in [29].

Example 4.2. Consider the system (2.1) with parameters as follows:

$$\begin{aligned}
A_1 &= \begin{bmatrix} -0.1 & 0 \\ 0.1 & -0.4 \end{bmatrix}, & A_{d1} &= \begin{bmatrix} 0.2 & 0 \\ -0.1 & -0.1 \end{bmatrix}, & B_1 &= \begin{bmatrix} 0 \\ 0.3 \end{bmatrix}, \\
E_1 &= \begin{bmatrix} -0.11 & 0.1 \\ 0 & 0.06 \end{bmatrix}, & E_{d1} &= \begin{bmatrix} 0.11 & 0.1 \\ 0 & 0.06 \end{bmatrix}, & G_1 &= \begin{bmatrix} -0.2 \\ 0.1 \end{bmatrix}, \\
C_1 &= \begin{bmatrix} -0.1 & 0.05 \\ 0.2 & 0.05 \end{bmatrix}, & C_{d1} &= \begin{bmatrix} -0.11 & 0.05 \\ 0.6 & 0.3 \end{bmatrix}, & D_1 &= \begin{bmatrix} 0.1 \\ -0.1 \end{bmatrix}, \\
M_{11} &= \begin{bmatrix} -0.03 \\ 0.02 \end{bmatrix}, & M_{21} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, & M_{31} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, & M_{41} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, \\
M_{51} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, & M_{61} &= \begin{bmatrix} 0.06 \\ 0.02 \end{bmatrix}, & M_{71} &= \begin{bmatrix} 0.01 \\ 0.05 \end{bmatrix}, & M_{81} &= \begin{bmatrix} 0.08 \\ -0.03 \end{bmatrix}, \\
N_{11} &= [0.02 \ 0.02], & N_{21} &= [0.02 \ 0.02], & N_{31} &= 0.01, \\
L_1 &= \begin{bmatrix} 0.1 & 0.1 \\ 0.12 & 0.1 \end{bmatrix}, & F_{11} &= 0.2 \sin(t), & F_{21} &= 0.3 \sin(t), & F_{31} &= 0.3 \sin(t), \\
A_2 &= \begin{bmatrix} -0.2 & -0.3 \\ 0.5 & 0.4 \end{bmatrix}, & A_{d2} &= \begin{bmatrix} 0.1 & -0.1 \\ 0.1 & 0.2 \end{bmatrix}, & B_2 &= \begin{bmatrix} 0 \\ -0.3 \end{bmatrix}, \\
E_2 &= \begin{bmatrix} -0.12 & -0.11 \\ 0.11 & 0.04 \end{bmatrix}, & E_{d2} &= \begin{bmatrix} 0.12 & -0.11 \\ 0.11 & 0.04 \end{bmatrix}, & G_2 &= \begin{bmatrix} -0.1 \\ 0.1 \end{bmatrix}, \\
C_2 &= \begin{bmatrix} -0.2 & -0.1 \\ 0.5 & 0.2 \end{bmatrix}, & C_{d2} &= \begin{bmatrix} -0.2 & -0.1 \\ 0.5 & 0.2 \end{bmatrix}, & D_2 &= \begin{bmatrix} 0.1 \\ 0.2 \end{bmatrix}, \\
M_{12} &= \begin{bmatrix} 0.07 \\ 0.01 \end{bmatrix}, & M_{22} &= \begin{bmatrix} 0.02 \\ 0.12 \end{bmatrix}, & M_{32} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, & M_{42} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, \\
M_{52} &= \begin{bmatrix} -0.03 \\ -0.02 \end{bmatrix}, & M_{62} &= \begin{bmatrix} 0.08 \\ 0.07 \end{bmatrix}, & M_{72} &= \begin{bmatrix} 0.04 \\ 0.07 \end{bmatrix}, & M_{82} &= \begin{bmatrix} 0.03 \\ -0.07 \end{bmatrix}, \\
N_{12} &= [0.01 \ 0.02], & N_{22} &= [0.01 \ 0.01], & N_{32} &= 0.01, \\
L_2 &= \begin{bmatrix} 0.2 & 0.19 \\ 0.1 & 0.1 \end{bmatrix}, & F_{12} &= 0.2 \sin(t), & F_{22} &= 0.3 \sin(t), & F_{32} &= 0.3 \sin(t)
\end{aligned} \tag{4.2}$$

and $v(t) = 10e^{-t}$, the membership functions are $h_1(\theta) = (1 - \sin(x_2))/2$, $h_2(\theta) = (1 - \sin(x_1))/2$, and the nonlinear vector function is $\phi(u) = (K_2 + K_1)/2 + (K_2 - K_1)/2 \sin(u)$, $K_1 = \text{diag}\{0.1, 0.3\}$, $K_2 = \text{diag}\{0.2, 0.4\}$.

Note that different τ_1 and τ_2 yield different γ_{\min} , and assume $\tau(k)$ satisfies $1 \leq \tau(k) \leq 5$.

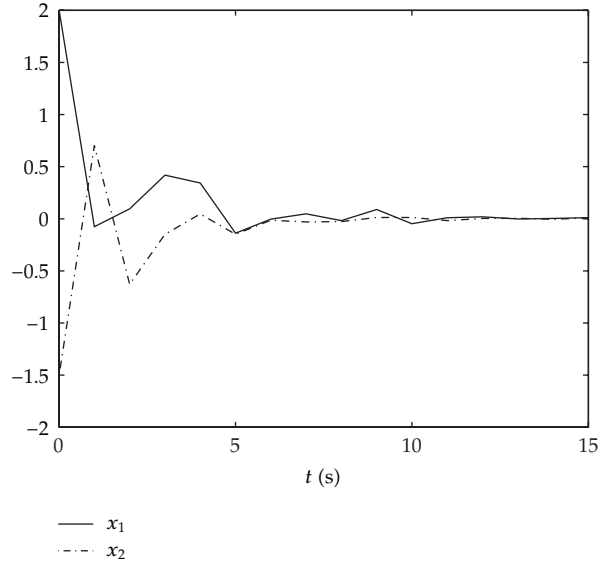


Figure 1: The state response $x(k)$.

By Theorem 3.1, the minimum achievable noise attenuation level is given by $\gamma_{\min} = 0.3474$ and we can obtain the corresponding filter parameters as follows:

$$\begin{aligned} A_{f1} &= \begin{bmatrix} -0.3876 & -0.6504 \\ 0.4967 & 0.8339 \end{bmatrix}, & B_{f1} &= \begin{bmatrix} 1.4538 & 0.8653 \\ -4.9933 & -1.7834 \end{bmatrix}, \\ A_{f2} &= \begin{bmatrix} -0.3600 & -0.6051 \\ 0.4593 & 0.7720 \end{bmatrix}, & B_{f2} &= \begin{bmatrix} 1.2017 & 0.7808 \\ -4.6656 & -1.6742 \end{bmatrix}. \end{aligned} \quad (4.3)$$

By Theorem 3.4, the minimum achievable noise attenuation level is given by $\gamma_{\min} = 0.3419$ and we can obtain the corresponding filter parameters as follows:

$$\begin{aligned} A_{f1} &= \begin{bmatrix} 0.5059 & 0.4421 \\ -0.7283 & -0.6452 \end{bmatrix}, & B_{f1} &= \begin{bmatrix} -0.3795 & 0.2226 \\ -2.1600 & -0.9045 \end{bmatrix}, \\ A_{f2} &= \begin{bmatrix} -0.5060 & -0.4420 \\ 0.7288 & 0.6429 \end{bmatrix}, & B_{f2} &= \begin{bmatrix} 1.6933 & 1.0131 \\ -3.1230 & -1.1367 \end{bmatrix}. \end{aligned} \quad (4.4)$$

By Theorem 3.5, the minimum achievable noise attenuation level is given by $\gamma_{\min} = 0.2563$ and we can obtain the corresponding filter parameters as follows:

$$\begin{aligned} A_{f1} &= \begin{bmatrix} -0.0002 & -0.0011 \\ 0.0016 & -0.0066 \end{bmatrix}, & B_{f1} &= \begin{bmatrix} -0.1708 & 0.3000 \\ -2.8286 & -1.0410 \end{bmatrix}, \\ A_{f2} &= \begin{bmatrix} -0.4489 & -0.3942 \\ 0.6516 & 0.5707 \end{bmatrix}, & B_{f2} &= \begin{bmatrix} 3.2336 & 1.6093 \\ -5.2772 & -1.9714 \end{bmatrix}. \end{aligned} \quad (4.5)$$

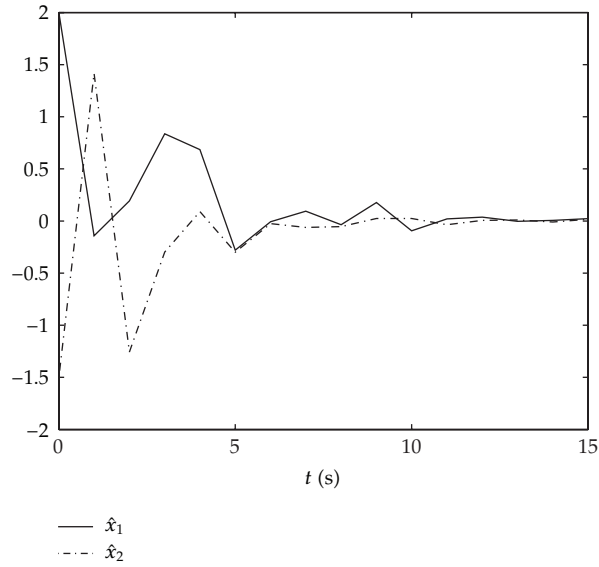


Figure 2: The estimation of filter $\hat{x}(k)$.

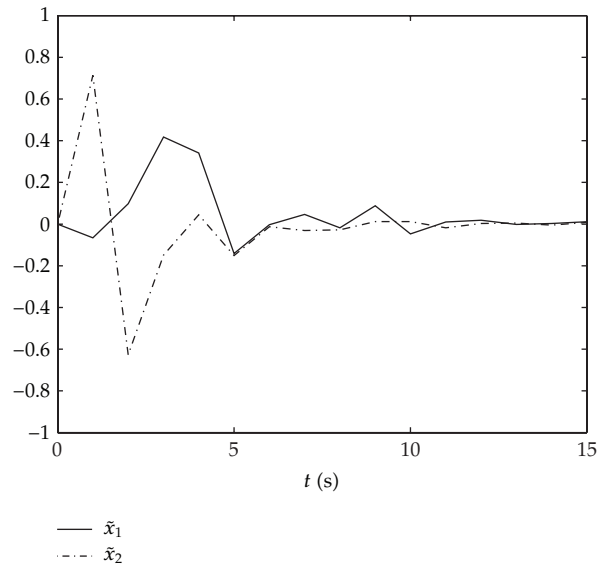


Figure 3: The error response $\tilde{x}(k)$.

The relaxation technique is adopted in this paper, so Theorems 3.4 and 3.5 relaxed the conservatism more than Theorem 3.1. There has been a decrease in the minimum achievable noise attenuation level γ_{\min} .

With the initial conditions $x(t)$ and $\hat{x}(t)$ are $[2 - 1.5]^T$ and $[2 - 1.5]^T$, respectively, for an appropriate initial interval. For given $\tau_1 = 1, \tau_2 = 5$ with $\gamma_{\min} = 0.2563$, according to Theorem 3.5, we apply the filter parameters mentioned previously to the system (2.1) and obtain the simulation results as in Figures 1–3. Figure 1 shows the state response $x(k)$

under the initial condition. Figure 2 shows the estimation of filter $\hat{x}(k)$. Figure 3 shows error response $\tilde{x}(k)$. From these simulation results, we can see that the designed H_∞ filter can stabilize the system (2.1) with sensor nonlinearities and time-varying delay.

5. Conclusions

In this paper the robust filtering problem for a class of uncertain discrete-time fuzzy stochastic systems with sensor nonlinearities and time-varying delay has been developed. A new type of Lyapunov-Krasovskii functional has been constructed to derive some sufficient conditions for the filters in terms of LMIs, which guarantees a prescribed H_∞ performance index for the filtering error system. A numerical example has shown the usefulness and effectiveness of the proposed filter design method.

References

- [1] T. Takagi and M. Sugeno, "Fuzzy identification of systems and its applications to modeling and control," *IEEE Transactions on Systems, Man and Cybernetics*, vol. 15, no. 1, pp. 116–132, 1985.
- [2] K. Tanaka and M. Sano, "Robust stabilization problem of fuzzy control systems and its application to backing up control of a truck-trailer," *IEEE Transactions on Fuzzy Systems*, vol. 2, no. 2, pp. 119–134, 1994.
- [3] T. Tanaka and H. O. Wang, *Fuzzy Control Systems Design and Analysis: A Linear Matrix Inequality Approach*, Wiley, New York, NY, USA, 2001.
- [4] K. Tanaka and M. Sugeno, "Stability analysis and design of fuzzy control systems," *Fuzzy Sets and Systems*, vol. 45, no. 2, pp. 135–156, 1992.
- [5] E. Kim and H. Lee, "New approaches to relaxed quadratic stability condition of fuzzy control systems," *IEEE Transactions on Fuzzy Systems*, vol. 8, no. 5, pp. 523–534, 2000.
- [6] C. Lin, Q.-G. Wang, and T. H. Lee, "A less conservative robust stability test for linear uncertain time-delay systems," *Institute of Electrical and Electronics Engineers*, vol. 51, no. 1, pp. 87–91, 2006.
- [7] H.-N. Wu, "Delay-dependent H_∞ fuzzy observer-based control for discrete-time nonlinear systems with state delay," *Fuzzy Sets and Systems*, vol. 159, no. 20, pp. 2696–2712, 2008.
- [8] S. Zhou, J. Lam, and A. Xue, " H_∞ filtering of discrete-time fuzzy systems via basis-dependent Lyapunov function approach," *Fuzzy Sets and Systems*, vol. 158, no. 2, pp. 180–193, 2007.
- [9] M. Chen, G. Feng, H. Ma, and G. Chen, "Delay-dependent H_∞ filter design for discrete-time fuzzy systems with time-varying delays," *IEEE Transactions on Fuzzy Systems*, vol. 17, no. 3, pp. 604–616, 2009.
- [10] V. B. Kolmanovskii and A. D. Myshkis, *Introduction to the Theory and Applications of Functional Differential Equations*, Kluwer, Dordrecht, The Netherlands, 1999.
- [11] M. Malek-Zavarei and M. Jamshidi, *Time-Delay Systems: Analysis, Optimization and Application*, North-Holland, Amsterdam, The Netherlands, 1987.
- [12] *Stability and Control of Time-Delay Systems*, vol. 228 of *Lecture Notes in Control and Information Sciences*, Springer, London, UK, 1998.
- [13] B. Zhang, S. Xu, and Y. Zou, "Improved stability criterion and its applications in delayed controller design for discrete-time systems," *Automatica*, vol. 44, no. 11, pp. 2963–2967, 2008.
- [14] X. Meng, J. Lam, B. Du, and H. Gao, "A delay-partitioning approach to the stability analysis of discrete-time systems," *Automatica*, vol. 46, no. 3, pp. 610–614, 2010.
- [15] H. Shao and Q.-L. Han, "New stability criteria for linear discrete-time systems with interval-like time-varying delays," *Institute of Electrical and Electronics Engineers*, vol. 56, no. 3, pp. 619–625, 2011.
- [16] J. Liu and J. Zhang, "Note on stability of discrete-time time-varying delay systems," *IET Control Theory & Applications*, vol. 6, no. 2, pp. 335–339, 2012.
- [17] Y. Su, B. Chen, C. Lin, and H. Zhang, "A new fuzzy H_∞ filter design for nonlinear continuous-time dynamic systems with time-varying delays," *Fuzzy Sets and Systems*, vol. 160, no. 24, pp. 3539–3549, 2009.
- [18] S. J. Huang, X. Q. He, and N. N. Zhang, "New Results on H_∞ Filter Design for Nonlinear Systems with Time Delay via T-S Fuzzy Models," *IEEE Transactions on Fuzzy Systems*, vol. 19, no. 1, pp. 193–199, 2011.

- [19] H. Yan, M. Q.-H. Meng, H. Zhang, and H. Shi, "Robust H_∞ exponential filtering for uncertain stochastic time-delay systems with Markovian switching and nonlinearities," *Applied Mathematics and Computation*, vol. 215, no. 12, pp. 4358–4369, 2010.
- [20] S. Wen, Z. Zeng, and T. Huang, "Reliable H_∞ filter design for a class of mixed-delay markovian jump systems with stochastic nonlinearities and multiplicative noises via delay-partitioning method," *International Journal of Control Automation, and Systems*, vol. 10, no. 4, pp. 711–720, 2012.
- [21] W. Yang, M. Liu, and P. Shi, " H_∞ filtering for nonlinear stochastic systems with sensor saturation, quantization and random packet losses," *Signal Processing*, vol. 92, no. 6, pp. 1387–1396, 2012.
- [22] H. Li and Y. Shi, "Robust H_∞ filtering for nonlinear stochastic systems with uncertainties and Markov delays," *Automatica*, vol. 48, no. 1, pp. 159–166, 2012.
- [23] L. Ma, F. Da, and K.-J. Zhang, "Exponential H_∞ filter design for discrete time-delay stochastic systems with Markovian jump parameters and missing measurements," *IEEE Transactions on Circuits and Systems*, vol. 58, no. 5, pp. 994–1007, 2011.
- [24] L. Wu and D. W. C. Ho, "Fuzzy filter design for Itô stochastic systems with application to sensor fault detection," *IEEE Transactions on Fuzzy Systems*, vol. 17, no. 1, pp. 233–242, 2009.
- [25] S. Xu, J. Lam, H. Gao, and Y. Zou, "Robust H_∞ filtering for uncertain discrete stochastic systems with time delays," *Circuits, Systems, and Signal Processing*, vol. 24, no. 6, pp. 753–770, 2005.
- [26] Z. Wang, J. Lam, and X. Liu, "Filtering for a class of nonlinear discrete-time stochastic systems with state delays," *Journal of Computational and Applied Mathematics*, vol. 201, no. 1, pp. 153–163, 2007.
- [27] Y. Liu, Z. Wang, and X. Liu, "Robust H_∞ filtering for discrete nonlinear stochastic systems with time-varying delay," *Journal of Mathematical Analysis and Applications*, vol. 341, no. 1, pp. 318–336, 2008.
- [28] B. Shen, Z. Wang, H. Shu, and G. Wei, " H_∞ filtering for nonlinear discrete-time stochastic systems with randomly varying sensor delays," *Automatica*, vol. 45, no. 4, pp. 1032–1037, 2009.
- [29] Y. Niu, D. W. C. Ho, and C. W. Li, "Filtering for discrete fuzzy stochastic systems with sensor nonlinearities," *IEEE Transactions on Fuzzy Systems*, vol. 18, no. 5, pp. 971–978, 2010.
- [30] H. Yan, H. Zhang, H. Shi, and M. Q. H. Meng, " H_∞ fuzzy filtering for discrete-time fuzzy stochastic systems with time-varying delay," in *Proceedings of the 29th Chinese Control Conference (CCC '10)*, pp. 5993–5998, Beijing, China, July 2010.
- [31] X.-G. Guo and G.-H. Yang, "Reliable H_∞ filter design for discrete-time systems with sector-bounded nonlinearities: an LMI optimization approach," *Acta Automatica Sinica*, vol. 35, no. 10, pp. 1347–1351, 2009.
- [32] B. Zhou, W. X. Zheng, Y.-M. Fu, and G.-R. Duan, " H_∞ filtering for linear continuous-time systems subject to sensor non-linearities," *IET Control Theory & Applications*, vol. 5, no. 16, pp. 1925–1937, 2011.
- [33] K. Gu, V. L. Kharitonov, and J. Chen, *Stability of Time-Delay Systems*, Birkhäuser, Boston, Mass, USA, 2003.
- [34] Y. Y. Wang, L. Xie, and C. E. de Souza, "Robust control of a class of uncertain nonlinear systems," *Systems & Control Letters*, vol. 19, no. 2, pp. 139–149, 1992.



Hindawi

Submit your manuscripts at
<http://www.hindawi.com>

